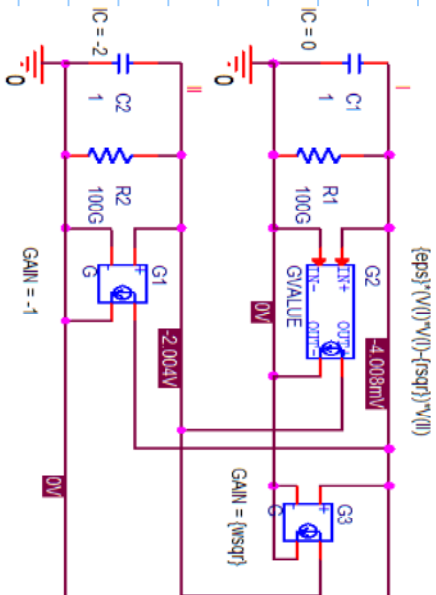


Homework 3. Also a good base paper topic may be Adiabatic Logic Circuits; a review paper is by V. I. Starosel'skii on the web.

Van der Pol Oscillator circuit. Passivity S matrix properties, all pass, circulator, loaded circulator



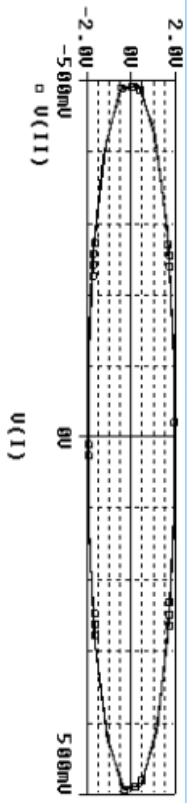
PARAMETERS:  
eps = 1  
rsqr = 1  
wsqr = 16

```
netlist:
* source VANDERPOL
C C1 1 0 1 IC=0 TC=0.0
C C2 1 0 1 IC=2 TC=0.0
G G1 1 0 1 0 -1
R R1 1 0 100G TC=0.0
R R2 1 0 100G TC=0.0
G G2 1 0 VALUE {
{eps*(V(1)-V(0)-f(sqrt)*V(1))}
G G3 1 0 1 0 {wsqr}
.PARAM eps=1 rsqr=1 wsqr=16
```

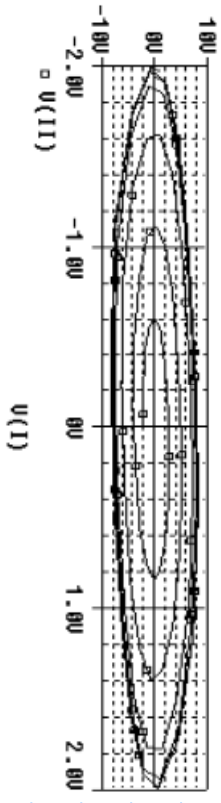
*ODE in circuits*

$$\ddot{x} + \epsilon(x^2 - 1)\dot{x} + \omega_0^2 x = 0, \quad x(0), \dot{x}(0) \quad \text{Van der Pol oscillator}$$

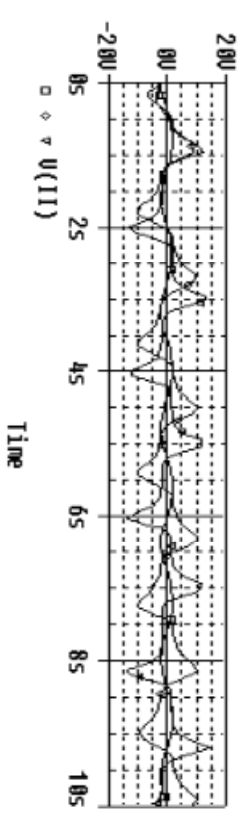
$$\begin{cases} \dot{x} = y \\ \dot{y} = \dot{x} = -\epsilon(x^2 - 1)y - \omega_0^2 x \end{cases} \quad \left. \begin{array}{l} \text{state variables} \\ \text{equations} \end{array} \right\}$$



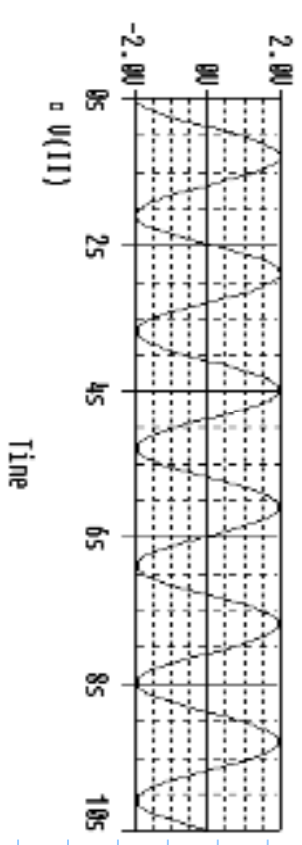
ICS: C1=>0,C2=>-2  
 eps=0



ICS: C1=>0,C2=>-2  
 eps=1



eps = 0.5 10

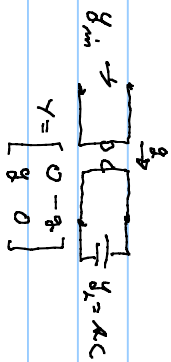


eps=0

Ex:  $\sum C$   $y_c = RC$ ,  $S_c(s) = (1+RCs)^{-1} (1-RC) = \frac{1-RC}{1+RC}$   $A = \sigma + j\omega$

$S_c(j\omega) = \frac{1-j\omega RC}{1+j\omega RC}$  ;  $|S_c(j\omega)| = \frac{\sqrt{1+(RC\omega)^2}}{\sqrt{1+(RC\omega)^2}} = 1$

$L = \frac{5}{g_1 + g_2}$



$y = \begin{bmatrix} 0 & -g_2 \\ g_1 & 0 \end{bmatrix}$

$y_{in}(s) = \frac{y_{11}g_1 + \Delta y}{g_{22} + y_L} = \frac{0 + g_1}{RC} = \frac{1}{RC} = \frac{1}{R(g/g_2)} = \frac{1}{R}$

$\Rightarrow \equiv \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \frac{g_1}{g_2} i_1$

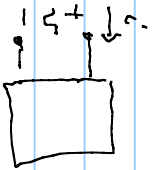


$i_1 = -g_1 v_1$   $-i_2 = -g_2 v_2$   $i_1 = g_1 v_1$   $i_2 = g_2 v_2 = g_2 \left( \frac{-i_1}{g_1} \right)$

$v_2 = T v_1$   
 $i_1 = -T i_2$

$P(t) = (v_1^T i_1 + v_2^T i_2) = v_1^T [i_1 + (T v_1)^T i_2] = 0$  *powerless in time*

Passivity  $P(t) = \int_{-\infty}^t P(\tau) d\tau$  If  $E(t) > 0$  for every  $t$ , including  $t = \infty$



$P(t) = v^T i = \frac{v^T i + i^T v}{2}$

$i = \sqrt{-1} = -j$   
complex conjugate

$E(\omega) = \int_{-\infty}^{\infty} v^T i d\tau \Rightarrow$  Parseval's theorem  $\int_{-\infty}^{\infty} v_c^T i_c dt = \int_{-\infty}^{\infty} \frac{1}{2} v_c^T I_c(j\omega) d\omega$

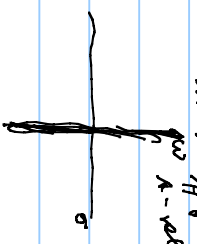
$\int_{-\infty}^{\infty} \frac{(v_c^T i_c + i_c^T v_c) d\tau}{2} = \int_{-\infty}^{\infty} \frac{V_c^T I_c(j\omega) + I_c^T(j\omega) V_c(j\omega) d\omega}{2} = \int_{-\infty}^{\infty} \frac{1}{2} V_c^T I_c(j\omega) d\omega$  *power equivalence*

$$= \int_{-\infty}^{\infty} \frac{(V(j\omega)^T Y(j\omega) V(j\omega)) + (V(j\omega)^T Y(j\omega) V(j\omega))}{2} \frac{d\omega}{2\pi}$$

$$= \int_{-\infty}^{\infty} \frac{V(j\omega)^T [Y(j\omega) + Y(j\omega)^T] V(j\omega)}{2} \frac{d\omega}{2\pi} = \int_{-\infty}^{\infty} V(j\omega)^T \underset{\text{Hermitian}}{Y(j\omega) + Y(j\omega)^T} \frac{d\omega}{2\pi}$$

for all  $V(j\omega)$  gives a power constraint  $Y_H(j\omega) = \frac{Y(j\omega) + Y(j\omega)^T}{2}$

If the circuit is passive with a  $Y(s)$   
 then  $Y_H(j\omega) \Rightarrow E(\omega) \geq 0 \Rightarrow Y_H(j\omega)$  is positive semi-definite



for  $\sigma > 0$  expect no singularity by stability

$$Y_H(s) = \frac{N(s)D^*(s)}{D(s)D^*(s)} = \frac{a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0}{d_n s^n + \dots + d_1 s + d_0}$$

$Y_H(s) = Y_H(j\omega)$  by analytic continuation  
 $j\omega = s$  we know for rational functions of  $\omega = \omega_j$

for a passive circuit

$Y(s)$  has no singularities  $\sigma > 0$

$Y(s) + Y^*(s)$  is positive semi-definite

positive real matrix

$Y(s)$  is real for real  $s, s = \sigma > 0$

If  $Y(s)$  doesn't exist & the circuit is passive then  $S(s)$  exists

Power for scattering  $P(t) = \int_{-\infty}^t \frac{v^T(i + i^T) v}{2} dt$ ;  $v = v^T + v^i$   
 $i = v^e - v^i$

$$v^T i = (v^T + v^i)^T (v^e - v^i)$$

$$v^T i = (v^T + v^i)^T (v^e - v^i) = v^T v^e - v^T v^i + v^i v^e - v^i v^i$$

$$v^T i = (v^T + v^i)^T (v^e - v^i) = v^T v^e - v^T v^i + v^i v^e - v^i v^i$$

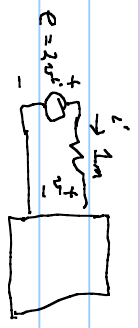
$$E(s) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} [u^T u^T - v^T v^T] dt = \int_{-\infty}^{\infty} (V^T V - V^T V) \frac{d\omega}{2\pi}$$

$$V(s) = S(s) V(s) = \int_{-\infty}^{\infty} V^T [I_n - S(s) S^T(s)] V \frac{d\omega}{2\pi}$$

for passive  $I_n - S(s) S^T(s)$  is positive semidefinite Bounded Real

now still no poles but in  $\sigma > 0$

$S(s)$  is real in  $\sigma > 0$



$$\int_{-\infty}^{\infty} u^T u^T v^T v^T \frac{d\omega}{2\pi} = \int_{-\infty}^{\infty} (V^T V + 2V^T I(s) I(s) V) \frac{d\omega}{2\pi}$$

$\Rightarrow \int_{-\infty}^{\infty} (V^T V + I^T I) \frac{d\omega}{2\pi} \geq 0$  &  $S$  is stable